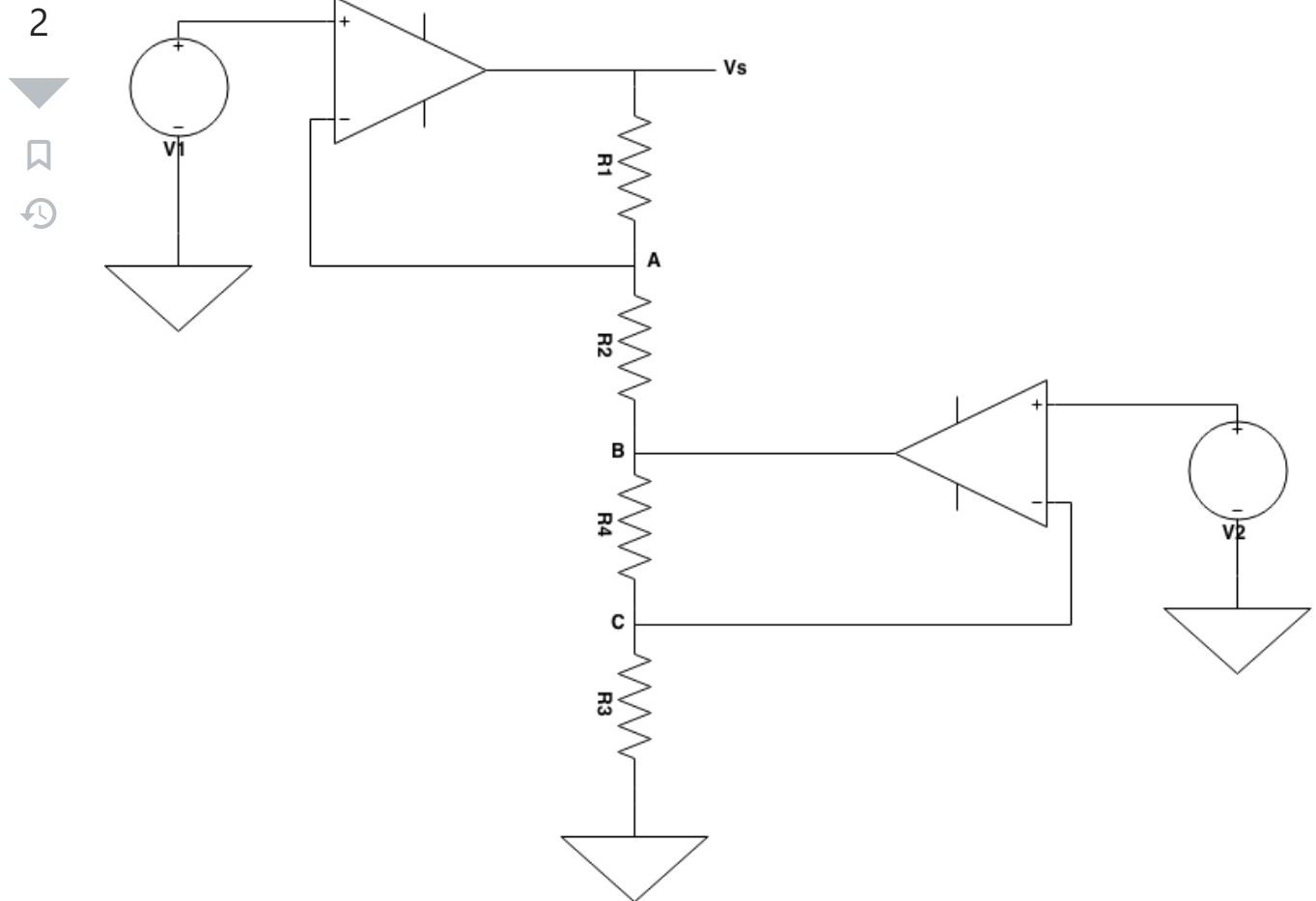


Transfer function of an op amp circuit

Asked 7 years, 9 months ago Modified 7 years, 9 months ago Viewed 2k times

▲ Let's suppose the following circuit:



I am trying to find a relation between V_s , V_1 and V_2 at this circuit. Please note that V_s is measured in relation to the ground. I have declared some extra points (A,B,C) on the circuit and using voltage divider laws and the properties of an ideal operational amplifier I came up with the following solution:

$$V_2 = V_c = V_B \frac{R_3}{R_3 + R_4} \Rightarrow V_B = \frac{V_2}{\frac{R_3}{R_3 + R_4}} \quad (1)$$

$$V_1 - V_B = (V_s - V_B) \frac{R_2}{R_1 + R_2} \quad (2)$$

$$(2) \xrightarrow{(1)} V_1 - \frac{V_2}{\frac{R_3}{R_3 + R_4}} = \left(V_s - \frac{V_2}{\frac{R_3}{R_3 + R_4}} \right) \frac{R_2}{R_1 + R_2}$$

$$\Rightarrow V_1 - \frac{V_2 \cdot (R_3 + R_4)}{R_3} = V_s \cdot \frac{R_2}{R_1 + R_2} - \frac{V_2 (R_3 + R_4)}{R_3} \cdot \frac{R_2}{R_1 + R_2}$$

$$\Rightarrow V_s = \frac{V_2 (R_3 + R_4)}{R_2} - V_1 - \frac{V_2 (R_3 + R_4)}{R_3} \cdot \frac{R_2}{R_1 + R_2}$$

Is this correct?

Another solution with a different result is the following:

$$V_c = V_2 \quad (1)$$

$$i_1 = \frac{V_c}{R_3} \stackrel{(1)}{=} \frac{V_2}{R_3} \quad (2)$$

$$i_2 = i_1 \stackrel{(2)}{=} \frac{V_2}{R_3} \quad (3)$$

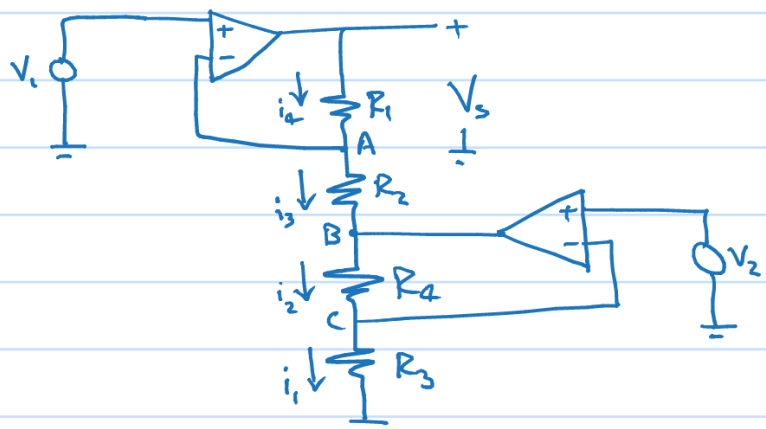
$$V_A = V_1 \quad (4)$$

$$i_3 = i_2 \stackrel{(3)}{=} \frac{V_2}{R_3} \quad (5)$$

$$i_4 = i_3 \quad (6)$$

$$i_4 = \frac{V_s - V_A}{R_1} \stackrel{(5,6)}{=} \frac{V_2}{R_3} \quad (4) \Rightarrow \frac{V_2}{R_3} = \frac{V_s - V_1}{R_1}$$

$$\Rightarrow V_2 \cdot \frac{R_1}{R_3} = V_s - V_1$$



$$\Rightarrow V_S = V_2 \cdot \frac{K_1}{R_3} + V_1$$

I am really confused.

operational-amplifier

ground

voltage-divider

transfer-function

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edited Dec 31, 2014 at 15:13

asked Dec 30, 2014 at 21:19



mgus

227 2 6 14

The second derivation is wrong: you write $i_3=i_2$ at node B, but you neglect the current from the output of the lower op-amp, which is not negligible in general. – Lorenzo Donati support Ukraine Dec 31, 2014 at 21:31

2 Answers

Sorted by:

Highest score (default)



Voltage division isn't a great approach to hang this on. Just build the output from the bottom up. You just need to know that current doesn't enter the input terminals of an op amp.

3



You know V_c , so you know the current through R_3 . That has to be the same as the current through R_4 , so now you know V_b . You also know V_a , so now you can calculate current through R_2 , which has to be the same as the current through R_1 , which would give you V_s .

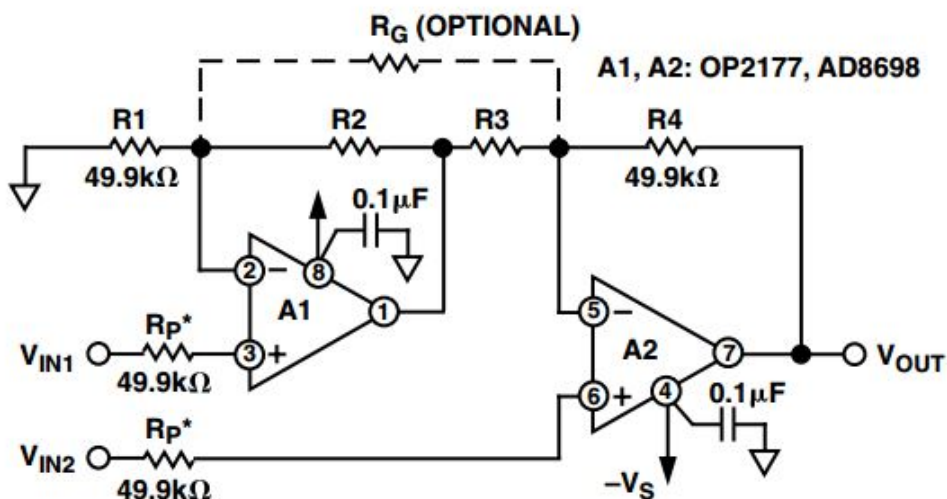


Looks a little like the input stage of an instrumentation amplifier, but there's an extra resistor.

CORRECTION This is a two op-amp instrumentation amplifier. Full discussion of the circuit at

[http://www.analog.com/static/imported-](http://www.analog.com/static/imported-files/design_handbooks/5812756674312778737Complete_In_Amp.pdf)

[files/design_handbooks/5812756674312778737Complete_In_Amp.pdf](http://www.analog.com/static/imported-files/design_handbooks/5812756674312778737Complete_In_Amp.pdf) on page 2-4.



$$V_{OUT} = (V_{IN2} - V_{IN1}) \left(1 + \frac{R_4}{R_3} + \frac{2R_4}{R_G} \right)$$

FOR $R_1 = R_4, R_2 = R_3$

*OPTIONAL INPUT PROTECTION RESISTOR FOR GAINS GREATER THAN 100 OR INPUT VOLTAGES EXCEEDING THE SUPPLY VOLTAGE

Figure 2-6. A 2-op amp in-amp circuit.

-- though the resistor numbers and inputs are not quite the same as what you use

Taking a shot at your derivation, starting from the line $i_2=i_1=V_2/R_3$, lets have a go at it

$$\begin{aligned} V_B &= V_2 + i_2 R_4 = V_2 + \frac{V_2 R_4}{R_3} \\ &= V_2 \left(1 + \frac{R_4}{R_3} \right) \\ i_3 &= \frac{V_A - V_B}{R_2} = \frac{V_1 - V_2 \left(1 + \frac{R_4}{R_3} \right)}{R_2} \end{aligned}$$

$$V_S = V_1 + i_4 R_1 = V_1 + \frac{V_1 R_1}{R_2} - V_2 \frac{R_1}{R_2} \left(1 + \frac{R_4}{R_3} \right)$$

or just shifting to make this look a bit more differential:

$$V_S = V_1 \left(1 + \frac{R_1}{R_2} \right) - V_2 \frac{R_1}{R_2} \left(1 + \frac{R_4}{R_3} \right)$$

That's a quick pass, and something feels wrong. Feel free to correct.

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edited Dec 31, 2014 at 18:47

answered Dec 30, 2014 at 22:32



Scott Seidman

28.3k 4 42 105

Thanks @acd -- I'm having a terrible time w/ a new autocorrect. – Scott Seidman Dec 30, 2014 at 22:50

I followed your advice and posted what I found (see above). But I ended up with a completely different result. Can you check it please? – mgus Dec 31, 2014 at 15:17

@KonstantinosKonstantinidis Will do. Editing Now – Scott Seidman Dec 31, 2014 at 18:20

Whew, TeX workout for someone who doesn't know it! – Scott Seidman Dec 31, 2014 at 18:39

I didn't notice anything wrong about your solution. But what I still don't understand is why voltage division I initially used gets us to a different solution. What is wrong about this? – mgus Dec 31, 2014 at 23:19



3



A clever circuit... beautiful and symmetrical (usually, $R_1 = R_2 = R_3 = R_4 = R$) like all the circuits of instrumentation amplifiers... It also gives a good opportunity to show how to make the unfamiliar circuits familiar - by dividing them into more well-known functional blocks instead of blindly analyzing them...

Structure. We can discern in this circuit of a *perfect instrumentation amplifier* two sub-circuits - an *imperfect (unbalanced) differential amplifier* (the top part consisting of the upper op-amp and the resistors R_1, R_2), and an ordinary *non-inverting amplifier* (the bottom part consisting of the lower op-amp and the resistors R_3, R_4).

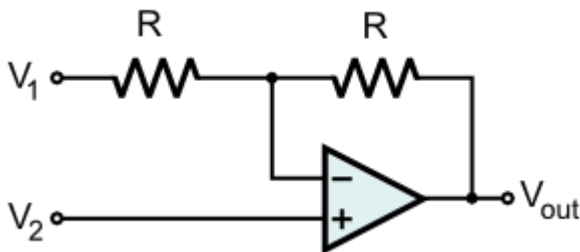
Analysis. Let's first consider the upper circuit part. With respect to V_1 , it is a non-inverting amplifier with gain of 2, and with regard to the lower input (from the side of the lower non-

inverting amplifier) - an inverting amplifier with gain of -1. As the lower non-inverting amplifier has a gain of 2, the two gains (inverting and non-inverting) of the imperfect differential amplifier are equalized... the two partial voltages superimpose and mutually neutralize at the op-amp output... and it becomes a perfect balanced differential (instrumentation) amplifier. From this perspective, the analysis is very simple:

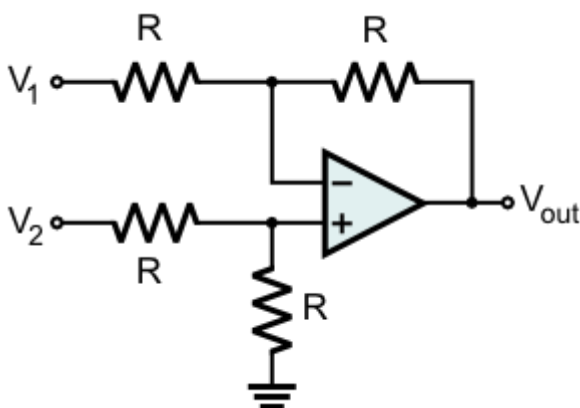
$$V_s = V_1 \cdot (R_1 + R_2) / R_2 - V_2 \cdot (R_3 + R_4) / R_3 \cdot R_1 / R_2 = 2V_1 - 2V_2 = 2(V_1 - V_2); \text{ Happy New Year!}$$

Philosophy. It is interesting to reveal the evolution of the op-amp differential amplifier to see where this circuit solution stays. I will use figurative (not generally accepted) names of the particular circuit solutions that are more meaningful. Also, to simplify this qualitative explanation, I suppose equal resistances (R).

1. **Unbalanced "differential amplifier"**. To make a differential amplifier, we need simply to subtract two input voltages. First we introduce a negative feedback by two resistors to obtain a fixed stable gain and then apply the two voltages to the inverting and non-inverting inputs of this "differential amplifier". Here the low resistance of the inverting input is a problem... but the big problem is that the two gains are not equal - the inverting gain (1) is less than the non-inverting gain (2). So, we have two choices to equalize them - to decrease (two times) the non-inverting gain or to increase (two times) the inverting gain. Let's consider them below...



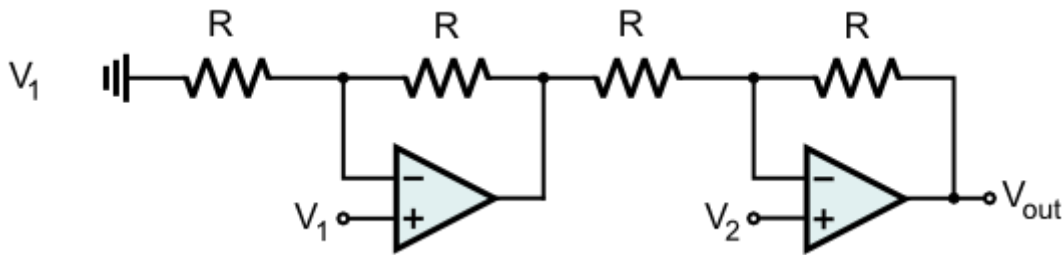
2. **Differential amplifier with non-inverting attenuation**. To decrease (two times) the non-inverting gain, we can connect a voltage divider (with two equal resistors) before the non-inverting input thus obtaining the classical 1-op-amp differential amplifier. The two gains are now equalized... but the high resistance of the non-inverting input is decreased...



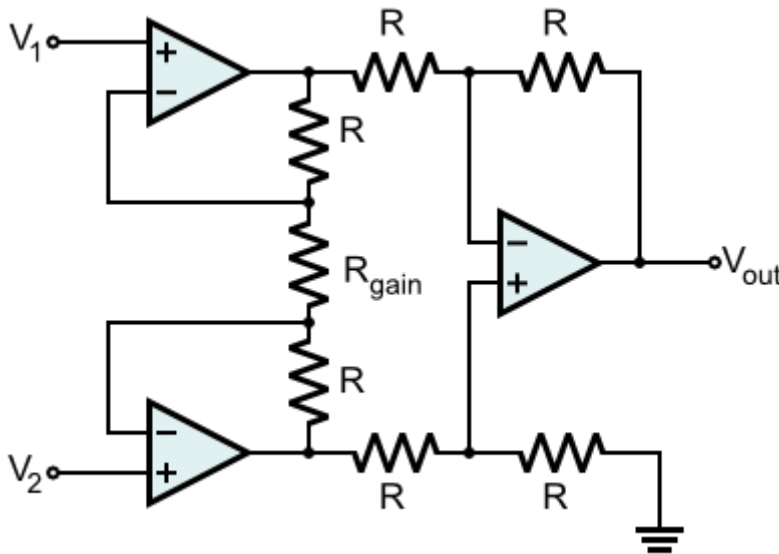
3. **Differential amplifier with inverting amplification**. With the same success we can

increase (two times) the inverting gain if we connect a non-inverting amplifier (with a gain of 2) before the inverting input (the discussed here solution). The two gains are equalized

again... and the both inputs have high resistance... It is a real 2-op-amp instrumentation amplifier.



4. **Buffered differential amplifier with non-inverting attenuation.** Finally, we can modify the classic 1-op-amp differential amplifier (case 2) by including non-inverting amplifiers before its inputs; this will solve the problems of the low input resistances. If we are inventive enough, we will combine the two lower resistors of the voltage dividers (inside the input non-inverting amplifiers) into one resistor (R_{gain}) that can regulate simultaneously both the input gains. Thus we will obtain the classic 3-op-amp instrumentation amplifier. It is interesting that there is a virtual ground in the middle point of R_{gain} ; it has replaced the real ground.



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edited Jun 11, 2020 at 15:10

answered Dec 31, 2014 at 10:28

 Community Bot
1


 Circuit fantasist
9,538 1 16 37

This doesn't seem right, given the circuit from the AD handbook I put up. – Scott Seidman Dec 31, 2014 at 18:44

Well, I will consider it... but now I have to go back to the New Year's table... the New Year comes here after three hours... Cheers! – Circuit fantasist Dec 31, 2014 at 19:34

Happy New Year!! – Scott Seidman Dec 31, 2014 at 20:13

@Scott Seidman, I have decided to start the new year by a fancy story about the evolution of the op-

operational amplifier - Transfer function of an op amp circuit - Electrical Engineering Stack Exchange
amp differential amplifier. BTW, as drawn, the AD circuit diagram is more appropriate for the intuitive understanding... – [Circuit fantasist](#) Jan 1, 2015 at 15:00 

- 1 @Circuitfantasist - See my reworking of your (so far) deleted without trace answer at address below. Seemed too good to let it go unremarked off into the void.
[electronics.stackexchange.com/questions/146442/...](https://electronics.stackexchange.com/questions/146442/) – [Russell McMahon](#) ♦ Jan 4, 2015 at 7:35
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